Rolling horizon for active fault detection

Miroslav Šimandl Ivo Punčochář Jakub Královec

Department of Cybernetics and Research Centre Data – Algorithms – Decision Making Faculty of Applied Sciences University of West Bohemia in Pilsen Czech Republic



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Passive and active fault detection Active detector design Information processing strategies Goals of the article

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Introduction

Passive and active fault detection

- Passive detector it provides only decision about faults [Willsky(1976), Basseville&Nikiforov(1993), etc.]
- Active detector in addition to decision it generates input signal to improve fault detection [Zhang(1989), Kerestecioğlu(1993), Campbell&Nikoukhah(2004)]



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Active detector design

Active detector design

- Generally, there are many methods to the detector design and it is not easy to sort them out
- There are only a few works on the active detector design, the availability of the future information is not considered and input signal design is based on an additional criterion
- Finally, there is not easy way to take into account real costs caused by the wrong decisions

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Passive and active fault detection Active detector design Information processing strategies Goals of the article

Introduction – cont'd

Information processing strategies (IPS's)

To sort out design methods the following strategies are considered in $[\check{S}imandl\&Herejt(2003)]$

- Open loop (OL) uses only an a priori information and possible future information is not considered
- Open loop feedback (OLF) uses an a priori information and available information up to current time, but the future information is not considered
- Closed loop (CL) uses an a priori information, available information up to current time and it is considered that the future information will be available too

Passive and active fault detection Active detector design Information processing strategies Goals of the article

Introduction – cont'd

Goals of the article

 Present a compact formulation of the active detector design in the multiple models framework and provide corresponding solution using CL IPS



Structutre of the active detector

• Propose a feasible suboptimal solution based on rolling horizon technique and compared it with OLF IPS and Pseudo Random Binary Sequence (PRBS) as input signal

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System description Active detector description and criterion

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Problem formulation

System description for $k \in \mathcal{T} = \{0, 1, \dots, F\}$

$$\begin{aligned} \mathbf{x}_{k+1} &= A(\theta_k)\mathbf{x}_k + B(\theta_k)u_k + G(\theta_k)\mathbf{w}_k \\ \mathbf{y}_k &= C(\theta_k)\mathbf{x}_k + H(\theta_k)\mathbf{v}_k \\ P_{i,j} &= P\left(\theta_{k+1} = j | \theta_k = i\right), \ i, j \in \mathcal{M} \end{aligned}$$

 \mathbf{x}_k unknown state; $\theta_k \in \mathcal{M} = \{1, 2, ..., N\}$ unknown index of the model; u_k input; \mathbf{y}_k output; \mathbf{w}_k and \mathbf{v}_k mutually independent white Gaussian noises; $A(\theta_k)$, $B(\theta_k)$, $G(\theta_k)$, $C(\theta_k)$, $H(\theta_k)$ known matrices; $P_{i,j}$ known transition probabilities; $p(\mathbf{x}_0)$ known Gaussian pdf; $P(\theta_0)$ known probability distribution

System description Active detector description and criterion

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Problem formulation – cont'd

Active detector description for $k \in T$

- Detector $d_k = \sigma_k \left(\mathbf{I}_0^k
 ight)$
- Input signal generator $u_k = \gamma_k \left(\mathsf{I}_0^k \right)$

with notation
$$\mathbf{I}_0^k = \left[{\mathbf{y}_0^k}^T, {u_0^{k-1}}^T \right]^T, \ \mathbf{y}_0^k = \left[{\mathbf{y}_0}^T, \dots, {\mathbf{y}_k}^T \right]^T$$

Criterion on finite detection horizon F

$$J\left(\sigma_{0}^{F},\gamma_{0}^{F}\right) = \mathrm{E}\left\{\sum_{k=0}^{F}L_{k}\left(\theta_{k},d_{k}\right)\right\} \to \min$$

So, the aim is to find functions $\sigma_0^{\rm F}$ and $\gamma_0^{\rm F}$ which minimize this criterion

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Active detector based on CL IPS Active detector based on OLF IPS Active detector based on rolling horizon technique State estimation in multiple models framework

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Active detector design

Active detector based on CL IPS

• Backward recursive equation with initial condition $V_{F+1}^* = 0$

$$V_{k}^{*}(\mathbf{I}_{0}^{k}) = \min_{d_{k}\in\mathcal{M}} \operatorname{E}\left\{L\left(\theta_{k}, d_{k}\right)|\mathbf{I}_{0}^{k}, d_{k}\right\}$$
$$+ \min_{u_{k}\in\mathcal{U}_{k}} \operatorname{E}\left\{V_{k+1}^{*}\left(\mathbf{I}_{0}^{k+1}\right)|\mathbf{I}_{0}^{k}, u_{k}\right\}, \ k = F, \dots, 0$$

- Detector $d_k^* = \arg \min_{d_k \in \mathcal{M}} \operatorname{E} \left\{ L(\theta_k, d_k) | \mathbf{I}_0^k, d_k \right\}$
- Input signal generator $u_{k}^{*} = \arg\min_{u_{k} \in \mathcal{U}_{k}} \mathbb{E}\left\{V_{k+1}^{*}\left(\mathbf{I}_{0}^{k+1}\right) | \mathbf{I}_{0}^{k}, u_{k}\right\}$

Active detector based on CL IPS Active detector based on OLF IPS Active detector based on rolling horizon technique

State estimation in multiple models framework

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Active detector design – cont'd

Active detector based on OLF IPS

• Modified aim is to solve OL problem at time step k for future time steps

$$\bar{J} = \min_{d_k^F, u_k^F} \mathbb{E} \left\{ \sum_{i=k}^F L\left(\theta_i, d_i\right) | \mathbf{I}_0^k, d_k^F, u_k^F \right\}$$

- Detector $d_k^{OLF} = \arg \min_{d_k \in \mathcal{M}} \mathbb{E} \left\{ L\left(\theta_k, d_k\right) | \mathbf{I}_0^k, d_k \right\}$
- Input signal generator is not determined ⇒ use of additional criterion or choose some probing signal (e.g. PRBS signal)

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Active detector design – cont'd

Active detector based on rolling horizon technique

• Aim is to numerically solve optimization problem for truncated horizon *I* with criterion

$$\tilde{J}\left(\sigma_{k}^{\bar{l}},\gamma_{k}^{\bar{l}}\right) = \mathbb{E}\left\{\sum_{i=k}^{\bar{l}}L\left(\theta_{i},d_{i}\right)\right\}, \ \bar{l} = \min\{k+l-1,F\}$$

and only $\sigma_k(\mathbf{I}_0^k)$, $\gamma_k(\mathbf{I}_0^k)$ are used

- Detector $d_k^{RH} = \arg\min_{d_k \in \mathcal{M}} \mathbb{E} \left\{ L(\theta_k, d_k) | \mathbf{I}_0^k, d_k \right\}$
- Input signal generator

$$u_{k}^{*} = \arg\min_{u_{k} \in \mathcal{U}_{k}} \operatorname{E}\left\{ \tilde{V}_{k+1}^{*} \left(\mathbf{I}_{0}^{k+1} \right) | \mathbf{I}_{0}^{k}, u_{k} \right\}$$

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State estimation in multiple models framework

Exact state estimation

- Given model sequence θ_0^k the state pdf $p(\mathbf{x}_k | \mathbf{I}_0^k, \theta_0^k)$ is computed using Kalman filter $\Rightarrow N^{k+1}$ filters are needed
- The state estimation is given as

$$p\left(\mathbf{x}_{k}|\mathbf{I}_{0}^{k}\right) = \sum_{\boldsymbol{\theta}_{0}^{k}} p(\mathbf{x}_{k}|\mathbf{I}_{0}^{k},\boldsymbol{\theta}_{0}^{k})P(\boldsymbol{\theta}_{0}^{k}|\mathbf{I}_{0}^{k})$$

$$P\left(\boldsymbol{\theta}_{k}|\mathbf{I}_{0}^{k}\right) = \sum_{\boldsymbol{\theta}_{0}^{k-1}} P(\boldsymbol{\theta}_{0}^{k}|\mathbf{I}_{0}^{k}) = \sum_{\boldsymbol{\theta}_{0}^{k-1}} \frac{p(\mathbf{y}_{k}|\mathbf{y}_{0}^{k-1},\mathbf{u}_{0}^{k-1},\boldsymbol{\theta}_{0}^{k})P(\boldsymbol{\theta}_{k}|\boldsymbol{\theta}_{k-1})P(\boldsymbol{\theta}_{0}^{k-1}|\mathbf{I}_{0}^{k-1})}{c}$$

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State estimation in multiple models framework - cont'd

Merging of the model sequences

• After each *h* time steps the model sequences which terminate in the same model are merged and resulting Gaussian mixture is replaced by single Gaussian distribution with the same first two moments

$$P(\theta_k | \mathbf{I}_0^k) = \sum_{\substack{\theta_{k-h}^{k-1}}} P(\theta_{k-h}^k | \mathbf{I}_0^k)$$
$$p(\mathbf{x}_k | \mathbf{I}_0^k, \theta_k) = \sum_{\substack{\theta_{k-h}^{k-1}}} \alpha(\theta_{k-h}^k) p(\mathbf{x}_k | \mathbf{I}_0^k, \theta_{k-h}^k) \approx p_A(\mathbf{x}_k | \mathbf{I}_0^k, \theta_k)$$
$$\alpha(\theta_{k-h}^k) = P(\theta_{k-h}^k | \mathbf{I}_0^k) / P(\theta_k | \mathbf{I}_0^k)$$

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State estimation in multiple models framework - cont'd

Replacement of exact pdf by single Gaussian distribution



Example 1 Example 2

Numerical examples

System description

$$\begin{aligned} \theta_{k} &= 1: \ x_{k+1} = 0.3x_{k} + u_{k} + \sqrt{0.25}w_{k} \\ y_{k} &= -2x_{k} + \sqrt{0.25}v_{k} \\ \theta_{k} &= 2: \ x_{k+1} = 0.5x_{k} + 1.5u_{k} + \sqrt{0.25}w_{k} \\ y_{k} &= 1.5x_{k} + \sqrt{0.25}v_{k} \end{aligned}$$

$$\begin{aligned} P_{i,j} &= \begin{bmatrix} 0.2 & 0.8 \\ 0.8 & 0.2 \end{bmatrix} & \theta_{k} = d_{k} \implies L(\theta_{k}, d_{k}) = 0 \\ \theta_{k} \neq d_{k} \implies L(\theta_{k}, d_{k}) = 1 \\ \theta_{0} &= 1) = P(\theta_{0} = 2) = 0.5 \\ w_{k} \sim \mathcal{N}\{0, 1\} & v_{k} \sim \mathcal{N}\{0, 1\} \end{aligned}$$

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Example 1 Example 2

Numerical examples – cont'd

Monte Carlo simulation results

• Parameters of the simulation F = 50, l = 2, h = 1, $U_k = \{-0.5, 0.5\}, k \in T$

	OLF+PRBS	RH	
$\mathrm{E}\{\hat{J}\}$	6.6142	2.7548	
$\operatorname{VAR}\{\hat{J}\}$	0.0083	0.0028	

• Improvement (6.6142 - 2.7548)/0.066142 = 58.35%

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Example 1 Example 2

Numerical examples – cont'd

Monte Carlo simulation results

• Parameters of the simulation F = 50, I = 2, h = 1, $\mathcal{U}_k = \{-\bar{u}, \bar{u}\}, \ k \in \mathcal{T}$

ū	0.1	0.3	0.5	1	3
$E\{\hat{J}^{OLF}\}$	18.5533	11.0350	6.6142	2.53	0.605
$\operatorname{VAR}\{\hat{J}^{OLF}\}$	0.0751	0.0977	0.0083	0.068	0.0018
$\mathrm{E}\{\hat{J}^{RH}\}$	17.0333	7.2633	2.7548	0.545	0.455
$\operatorname{VAR}\{\hat{J}^{RH}\}$	0.2687	0.1928	0.0028	0.0044	0.0024
D	1.52	3.7717	3.8594	1.985	0.15

$$D = \mathrm{E}\{\hat{J}^{OLF}\} - \mathrm{E}\{\hat{J}^{RH}\}$$

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Example 1 Example 2

Numerical examples - cont'd



Example 1 Example 2

Numerical examples - cont'd

System description

$$\begin{aligned} \theta_k &= 1: \ x_{k+1} = 0.98x_k + 2.6u_k + \sqrt{0.02}w_k \\ y_k &= 4.5x_k + \sqrt{0.3}v_k \\ \theta_k &= 2: \ x_{k+1} = 0.99x_k + 2.78u_k + \sqrt{0.02}w_k \\ y_k &= 4.4x_k + \sqrt{0.3}v_k \end{aligned}$$
$$\begin{aligned} P_{i,j} &= \begin{bmatrix} 0.99 & 0.01 \\ 0.01 & 0.99 \end{bmatrix} \qquad \begin{array}{l} \theta_k &= d_k \implies L(\theta_k, d_k) = 0 \\ \theta_k &\neq d_k \implies L(\theta_k, d_k) = 1 \\ \theta_0 &= 1) = P(\theta_0 = 2) = 0.5 \\ w_k \sim \mathcal{N}\{0, 1\} \qquad v_k \sim \mathcal{N}\{0, 1\} \end{aligned}$$

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Example 1 Example 2

Numerical examples – cont'd

Monte Carlo simulation results

• Parameters of the simulation F = 50, l = 2, h = 1, $U_k = \{-0.5, 0.5\}, k \in T$

	OLF+PRBS	RH	
$\mathrm{E}\{\hat{J}\}$	13.5520	4.4470	
$VAR{\hat{J}}$	0.8917	0.1110	

• Improvement (13.5520 - 4.4470)/0.13552 = 67.19%

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Conclusion remarks

Conclusion remarks

- The compact formulation of the fault detection problem was shown
- The stress was laid on the active fault detection problem and the suboptimal feasible solution was proposed
- Utilization of the closed loop information processing strategy leads to the lower value of the criterion
- Extensive MC simulations confirm that rolling horizon technique can provides better results in comparison with open loop feedback information processing strategy
- It was shown that possible improvement is highly dependent on considered input signal

Some comparison

Some comparison

- The main difficulty of some comparison consist in different design aims
- It is possible to make a very rough comparison with approach presented in [Zhang(1989)], where the detector is based on SPRT test and the input signal is designed to minimize ASN (Average Sampling Number)

	RH	Zhang
ASN	6.0486	13.45

 Unfortunately, the detector based on criterion minimization does not guarantee any limits on false alarm probability and missed alarm probability

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